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**APPLICATION  
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**TITLE: HIGH POWER AMPLIFIER AND CHASSIS**

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**SPECIFICATION**

## HIGH POWER AMPLIFIER AND CHASSIS

### Related Applications

This application claims the filing benefit and priority of U.S. Provisional Application entitled "A Low Cost Compact High Power Amplifier and Chassis," Serial No. 60/271,967, filed February 28, 2001, and incorporates that application by reference herein in its entirety.

### Field of the Invention

The present invention relates generally to high power amplifiers and specifically to a high power amplifier assembly having a single printed circuit board capable of housing and isolating a plurality of subcircuits.

### Background of the Invention

Modern high power amplifiers, such as those for cellular communications, do more than simply amplify a signal. Such power amplifiers also handle other tasks as well. For example, high power amplifiers often must also condition or change input prime power from one voltage to another voltage for proper operation of the internal

amplification circuits. Furthermore, such amplifiers may be required to translate an incoming frequency to a different frequency.

Another operation many high power amplifiers perform is the monitoring of internal circuit functionality and the transmission of that monitoring information 5 across an external data bus. Still further, high power amplifiers may be required to receive external operating instructions as well.

Currently, to properly provide such disparate operations and functions, existing high power amplifiers include numerous, separate printed circuit assemblies, which are often mounted and housed individually. For example, a separate board 10 assembly for each function is common in existing amplifier designs. The assemblies both contain and electrically isolate the various individual circuit assemblies. In such designs, each function is handled by its own separate printed circuit assembly which is mounted its own housing. With each function handled by stand-alone circuits, the various circuit assemblies then must be individually assembled and ultimately joined 15 into a single, completed assembly.

As may be appreciated, current power amplifier designs result in multiplication of numerous manufacturing steps, thus increasing production costs. Furthermore, the separate handling required for each separate circuit adds additional costs. For example, each separate board assembly for each of the various functions must be 20 separately kitted, assembled, cleaned and inspected. Often, several of these subassemblies are also tested prior to top level assembly and must be again retested once top level assembly occurs.

Additionally, discrete circuit connection components must be used to interconnect the various separate subassemblies and transmit signals from one 25 subassembly to the next. The interconnection components and assembly further

add complexity and cost to the design and production of the amplifier. The reliability of the overall structure is reduced as well, due to the multiple individual components for interconnection.

Still further, the size of such amplifiers is dictated by the amount of space  
5 necessary to incorporate and assemble all of the individually housed subassemblies together. Therefore, existing amplifiers often do not efficiently address size and space issues.

There exists a need for a high power amplifier which reduces the component manufacturing and production costs while increasing the reliability and maintaining  
10 the desired isolation of the high power amplifier within a compact assembly size.

#### **Brief Description of the Drawings**

The accompanying drawings, which are incorporated in and constitute a part of this specification, illustrate embodiments of the invention and, together with a  
15 general description of the invention given above, and the detailed description of the embodiments given below, serve to explain the principles of the invention.

Figure 1A is a top view of a high power amplifier in accordance with one embodiment of the present invention;

Figure 1B is a schematic circuit diagram of a high power amplifier of one  
20 embodiment of the present invention;

Figure 2A is a perspective view of a top side of a chassis body according to one embodiment of the present invention.

Figure 2B is a perspective view of a bottom side of a chassis body according to one embodiment of the present invention.

Figure 3A is a perspective view of a top side of a chassis lid according to one embodiment of the present invention.

Figure 3B is a perspective view of a bottom side of a chassis lid according to one embodiment of the present invention.

5       Figure 4A is a side view in partial cross-section of a portion of a printed power amplifier circuit board according to one embodiment of the present invention.

Figure 4B is a side view in partial cross-section of another portion of a printed power amplifier circuit board according to one embodiment of the present invention.

10      Detailed Description of Embodiments of the Invention

Referring to Figure 1A, a power amplifier assembly 10 of one embodiment of the present invention is shown and includes a single printed circuit board 12 and a chassis, including a chassis base or body 14, and a lid or lid structure 54. The chassis body, in combination with a lid structure or lid 54 (Figures. 3A and 3B) is 15 configured so that a single printed circuit board 12 contains multiple and functionally different subcircuits, which are shielded from undue electromagnetic (EM) and/or radio frequency (RF) interference and noise. Undue interference and noise is defined as interference and noise above a level currently tolerated in assemblies using multiple printed circuit boards, or noise which unacceptably reduces the ability 20 of the power amplifier assembly 10 to function properly. In addition, the chassis formed by the lid 54 and the chassis body 14 provides a single ground reference for all signals in the power amplifier assembly 10, thereby reducing the number of discrete wiring components. Still further, the entire assembly is housed within a single chassis or housing.

In one embodiment, the power amplifier assembly is adapted to amplify radio frequency (RF) signals. Currently, as discussed above in the Background Section, high power amplifiers are also required to do a number of additional functions. As noted, these functions are traditionally performed in existing amplifier designs by

5 using a series of separate printed circuit boards, each having specialized components and housing requirements. However, by using the power amplifier assembly 10 of the present invention, shielding is provided in component selection and lid compartmentalization to reduce the interference, allowing for the use of a single printed circuit board and a single housing instead of multiple printed circuit

10 boards and separate housings. In accordance with aspects of the invention, this is accomplished through a combination of the arrangement of the components on the printed circuit board 12, the use of ground planes within the printed circuit board 12, and the use of extensive shielding in the lid and common grounding in the housing.

15 These various aspects and other aspects are discussed below in reference to Figures 2-5.

Figure 1A illustrates a top view of a printed circuit board 12 inside a chassis body 14 with circuit components extending upwardly from a first conductive layer 16 of the printed circuit board 12 (other layers of the printed circuit board are visible in Figures 4A and 4B). In one embodiment, the various circuit components are

20 arranged into groups or subcircuits which are positioned on the board 12 and perform various functions for the power amplifier assembly 10. These subcircuits are separated by the dotted lines indicated as "A" in Figures 1A and 1B, which form various ground isolation paths.

In one embodiment of the present invention, the printed circuit board 12

25 includes a power supply subcircuit 18, an upconverter subcircuit 20, a monitor and

control subcircuit 22, a high power gain subcircuit 24, and a low power gain subcircuit component 26. It is also contemplated that the high power gain subcircuit 24 be on the printed circuit board with any combination of these circuits. Although a particular layout and arrangement of subcircuits is shown in Figure 1A, it is to be  
5 understood that the principles of the present invention may be used with alternative subcircuit layouts.

The power supply subcircuit 18 acts to provide the other subcircuits on the printed circuit board 12 with power. The power supply subcircuit 18 is separated from the other components by an isolation path 28. An isolation path may be defined  
10 as a printed strip of grounded circuit, which is dotted by a plurality of plated-through vias 30, as discussed further below. The isolation path 28 acts in conjunction with the internal chassis lid walls 64 (shown in Figure 3B) to electrically isolate the circuitry of each subcircuit on the printed circuit board 12 from EM and/or RF interference from the other subcircuit components.  
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When the power supply subcircuit 18 is placed on the printed circuit board with the high power gain subcircuit 24, it is desirable that the two components be placed away from each other in order to minimize magnetic interaction between the subcircuits that penetrates the chassis. In the embodiment shown in Figure 1A, the power supply subcircuit 18 and the high power gain subcircuit 24 are shown with the  
20 low power gain subcircuit 26 positioned between them. In an embodiment where the power supply subcircuit 18 and the high power gain subcircuit 24 are the only two subcircuits to be included on a single printed circuit board 12, the two subcircuits may have to be isolated by the inclusion of "Mu-metal," a ferrous alloy that magnetic fields cannot penetrate. Desirably, switching noise from the power supply subcircuit

18 should not show up as side bands on the output spectrum greater than -65 dB relative to the main signal.

The upconverter subcircuit 20 incorporates circuitry to convert RF signals from a lower frequency to a higher frequency. In one embodiment, the upconverter  
5 subcircuit 20 upconverts a signal between about 870 MHz and about 890 MHz to a signal between about 1930 MHz and about 1990 MHz signal. That signal is then amplified from about 125 microwatts to about 400 microwatts. The upconverter subcircuit 20 is also preferably separated from the power supply subcircuit 18 by the isolation path 28 and the internal chassis lid walls 64.

10 The monitor and control subcircuit 22 monitors the output of the other subcircuits to ensure that the subcircuits are operating correctly. The monitor and control subcircuit 22 may compare the other subcircuits' monitoring circuit signals against preset values. If the outputs do not match the preset values, the monitor and control subcircuit 22 may adjust signal levels within other subcircuits and/or within  
15 the monitor and control subcircuit until the output value and preset value match, for example, utilizing software closed circuit feedback in one embodiment. In one embodiment, the monitor and control subcircuit 22 is also separated from the other subcircuits by the isolation path 28 in conjunction with the internal chassis lid walls 64.

20 The low power gain subcircuit 26 performs a first amplification step, to a level of about 400 milliwatts, on signals coming from the upconverter. The high power gain subcircuit 24 performs a second amplification step to about 60 watts, which is delivered to an output connector of the power amplifier assembly 10. The high power gain subcircuit 24 is also surrounded by the isolation path 28 and the internal  
25 chassis lid walls 64. An output power detector subcircuit 32 serves to monitor the

output of the printed circuit board 12. The output power detector subcircuit 32 forwards this output power information to the monitor and control subcircuit 22. In one embodiment, the output power detector subcircuit 32 includes an integrated circuit manufactured by Analog Devices (part number AD8314).

5 By placing the power supply subcircuit 18, the upconverter subcircuit 20, the monitor and control subcircuit 22, the high power gain subcircuit 24, and the low power gain subcircuit 26 on a single printed circuit board 12, a variety of efficiencies are achieved. For example, excess input and output circuitry and connectors are eliminated because the connections between the various subcircuits may be made  
10 on or through the printed circuit board 12 itself. Further, in accordance with one aspect of the invention and using the isolation path 28, lid walls 64, and the EM and RF isolation techniques of the present invention, separate housings for each subcircuitry are not necessary. This represents significant material and assembly cost savings by eliminating multiple housings. EM and RF isolation equal to that  
15 achieved by using separate subcircuits within separate housings may be achieved using the present invention. Because only one unit has to be manufactured and tested, manufacturing costs are significantly reduced and product reliability is increased utilizing the present invention.

Turning now to Figure 1B, a block diagram of a high power amplifier circuit for  
20 use in one embodiment of the present invention is shown. The subcircuit components are numbered and lettered to indicate their respective subcircuits. Various values are set forth herein for embodiments of the invention, but are not meant to be limiting. The upconverter subcircuit 20 includes an input 20a for accepting a signal between about 869 and about 894 MHz at a power of about  
25 10.8dBm. An excessive input level attenuator 20b receives signals from the monitor

and control subcircuit 22 to keep power levels at or below a nominal 9dBm. The attenuator's nominal attenuation is 2dB. A 15dB coupler 20c sends an input power diagnostic signal through an integrated input power detector circuit 20e to the monitor and control subcircuit 22 for automatic gain control. Figure 1B shows a  
5 termination 20d, and in one embodiment a 50 ohm termination is used. A pre-mixer attenuator 20f attenuates the signal before the signal enters an upconverter mixer 20g and provides the upconverter mixer 20g with a good 50 ohm source. In one embodiment, the pre-mixer attenuator attenuates approximately -11dB.

The upconverter mixer 20g accepts the attenuated Intermediate Frequency (IF) signal and an LO signal from a phase-locked loop (PLL) synthesizer 20h to produce a radio frequency (RF) output. The PLL synthesizer 20h receives a synthesizer control signal from assembly connector J3 and forwards a "phase lock" monitor signal to the monitor and control subcircuit 22. The RF output from the upconverter mixer 20g is attenuated by a post-mixer attenuator 20i which serves to  
15 provide a 50 ohm load to the upconverter mixer 20g. The attenuated signal feeds an upconverter gain stage 20j. In one embodiment, the post-mixer attenuator 20i attenuates at about -5dB and the upconverter gain stage 20j is circuit model number MGA82563 manufactured by Agilent Corporation. Following the upconverter gain stage 20j, the signal enters an upconverter band pass filter 20k. In one embodiment,  
20 the upconverter band pass filter 20k has a center frequency of about 1960 MHz and a band width of about 60 MHz, and attenuates in-band signals at about -3 dB. Of course, other suitable components with different operating parameters may be utilized without deviating from the present invention.

The output signal from the upconverter subcircuit 20 next enters the low  
25 power gain subcircuit 26, where it is attenuated by an automatic gain control and

temperature compensation attenuator 26a, which attenuates in response to signals from the monitor and control subcircuit 22 at a nominal -4 dB. The attenuated signal then enters a first low power gain stage 26b. In one embodiment, the first low power gain stage is circuit model number MGA82563 manufactured by Agilent Corporation  
5 and it amplifies the incoming signal at about +13 dB. Next, the signal enters a band pass filter 26c. In one embodiment, the low power gain band pass filter 26c has a center frequency of about 1960 MHz, a band width of about 60 MHz, and attenuates in-band at about -3 dB.

The signal then enters a second low power gain stage 26d. In one embodiment, the second low power gain stage 26d is circuit model number FLU10XM manufactured by the Fujitsu Corporation and it amplifies the incoming signal at about +13 dB. A temperature sensor circuit 26e may send a temperature signal corresponding to the temperature of the low power gain subcircuit to the monitor and control subcircuit 22. In one embodiment, the temperature sensor circuit 26e is circuit model number LM50 manufactured by National Semiconductor.  
10 Following the second low power gain stage 26d, the signal enters a third low power gain stage 26f. In one embodiment, the third low power gain stage 26f is circuit model number FLU35XM manufactured by the Fujitsu Corporation, and it amplifies the incoming signal at about +11 dB. In one embodiment, a low power gain circuit  
15 isolator 26g is provided before the output of the low power gain subcircuit 26 in order to provide a 50 ohm load to the third low power gain stage 26f, as well as a 50 ohm source to the high power gain subcircuit 24.

The signal exiting the low power gain subcircuit 26 enters the high power gain subcircuit 24 and is amplified by a first high power gain stage 24a. In one embodiment, the first high power gain stage 24a is circuit model number FLL400IP-2  
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manufactured by the Fujitsu Corporation, and it amplifies the incoming signal at about +10 dB. The output of the first high power gain stage 24a is input into second and third high power gain stages 24b and 24c, which are coupled together in parallel. In one embodiment, the second and third high power gain stages 24b and

5 24c are circuit model number TPM1919-60 manufactured by the Toshiba Corporation each amplifying the incoming signal at about +12 dB, including losses in a power splitter and combiner. The output signal from the second and third high power gain stages 24b and 24c may be detected through the use of a 28 dB detection coupler 24d, with an associated detection coupler attenuator. In one

10 embodiment, the detection coupler is provided with attenuators which attenuate at about 41 dB. The detection coupler 24d is shown with a 50 ohm termination 24e. An output detector circuit 24f (which is part of the output power detector subcircuit 32) reads the power at this point and sends an output power signal to the monitor and control subcircuit 22. The output power detector circuit 24 may be preceded by

15 attenuators, and in one embodiment, a first output power attenuator attenuates at about -5dB, and a second output power attenuator attenuates at about -36dB. The signal from the second and third high power gain stages 24b and 24c passes through an isolator 24g before exiting the circuit at the output 24h.

Figure1B also shows a block diagram of the power supply subcircuit 18 and

20 the monitor and control subcircuit 22, with arrows and lines showing various inputs and outputs of these two subcircuits. The power supply subcircuit 18 includes power supply inputs 18a, providing between about 19 DC volts and about 30 DC volts, a continuous +15V DC output 18b, a switched +12V DC output 18c (controlled by the monitor and control subcircuit 22), a continuous +5V DC output 18d, and a

continuous -3V DC output 18e. The +5V DC output 18d is the power input into the monitor and control subcircuit 22.

The monitor and control subcircuit 22 includes a temperature signal input 22a, which receives signals from the temperature sensor 26e, an output detector signal 5 input 22b, which receives signals from the output detector circuit 24f, and an input detector subcircuit 22c, which receives signals from the input power detector circuit 20e. The monitor and control subcircuit 22 further includes an automatic gain control and temperature control attenuator output 22d, which sends signals to the automatic 10 gain control and temperature compensation attenuator 26a; an excessive input control output 22e, which sends signals to the excessive input level attenuator 20b; a fault output 22f, which sends fault signals to connector J3; an LED control output 22g, which controls LED indicators; and a phase lock detector input 22h, which receives signals from the PLL synthesizer 20h. Bias monitors 22i may also be provided in the monitor and control subcircuit 22, to monitor voltages within the 15 power supply subcircuit 18. The monitor and control circuit 22 may disable the amplifier 10 when a bias monitor goes out of a predetermined range.

Turning now to Figure 2A, a perspective view of a top side 34 of a conductive chassis body 14 according to one embodiment of the present invention is shown. In one embodiment, the chassis body 14 is cast out of aluminum and is electrically 20 conductive. Other materials may be used for the chassis body 14, such as copper, bronze or brass. Further, the chassis body 14 may be formed using other methods, such as milling. In accordance with one aspect of the present invention, a number of structures are formed in the chassis body 14. For example, a support boss 36 is provided to support a toroidal inductor, such as may be included in the power supply 25 subcircuit 18. Further, a variety of channels or depressions may be formed along the

conductive top side 34 of the chassis body. One such channel is a power detector coupling channel 38 for a coupled output signal feeding the output power detector subcircuit 32. Another channel is a monitor and control subcircuit pocket 40, which provides clearance for many signal paths in the monitor and control subcircuit 22, as well as for prime power feeding the power supply subcircuit 18. The chassis body is grounded and provides a good ground reference for the amplifier assembly 10.

A number of circuit component pockets 42 may also be provided along the top side 34 of the chassis body 14. In one embodiment, the circuit component pockets 42 are provided for flange-mounted power transistors extending downwardly from the printed circuit board 12 past its mating surface with the chassis body 14. In one embodiment, the circuit component pockets are about 13 mils deep from a main surface 44 of the chassis body top side 34. External connector paths 46 and 48 are provided to allow signal connections into and out of the power amplifier assembly 10. In one embodiment, external connector paths 46 and 48 are inputs, with additional pins provided on a connector into the connector path 46 for outputs. Other RF signal paths may be provided, such as a 15 MHz reference input for the upconverter subcircuit 20, a 900 MHz input, and a 1900 MHz output. Cooling fins 50 may also be provided as part of the chassis body 14 to dissipate heat from the power supply subcircuit 18, the high power gain subcircuit 24, and the other subcircuits on the printed circuit board 12. The cooling fins 50 are more clearly seen in Figure 2B, which also shows cooling fin casting ejector pads 52, for use when pushing a cast chassis body 14 out of its die.

The chassis body 14 works in conjunction with a chassis lid 54, (shown in Figure 3A, which provides a perspective view of a top side 55 of the chassis lid 54) to provide protection and EM and RF shielding to the various subcircuits of the

power amplifier assembly 10, in accordance with one aspect of the present invention. The chassis body is provided with a number of chassis lid attachment holes 56 for a secure connection between the chassis lid 54 and the chassis body 14. In one embodiment, the chassis lid attachment holes 56 accommodate thread-forming attachment screws which fit through edge attachment through holes 58 provided in the chassis lid 54. Central attachment through holes 60 may also be provided in the chassis lid 54 to support the chassis lid 54 and to ensure that gasketing material dispensed on the internal chassis lid walls 64 stays in contact with the printed circuit board 12 for desirable electrical isolation between the subcircuits.

Turning now to Figure 3B, a perspective view of a bottom side 62 of the chassis lid 54 illustrates a number of subcircuit cavities adapted to surround and protect the various subcircuits on the printed circuit board 12 and to provide EM and RF isolation between the subcircuits. The subcircuit cavities are formed by internal chassis lid walls 64 in conjunction with side walls 67 of the chassis body 14 (Figure 1A). The walls 64 generally extend downwardly from the lid. The internal chassis lid walls 64 are provided with various chassis lid wall pathways 66 wherever connections between subcircuits are made along a top side 68 (Figure 1A) of the printed circuit board 12. Subcircuit cavities are provided by the walls to electrically isolate subcircuits which require isolation, and are numbered in agreement with the numbering of their associated subcircuits that they isolate. The shown subcircuit cavities in the disclosed embodiment include a power supply subcircuit cavity 18', an upconverter subcircuit cavity 20', a monitor and control subcircuit cavity 22', a high power gain subcircuit cavity 24', a low power gain subcircuit cavity 26', and an output power detector subcircuit cavity 32'.

When the amplifier of the present invention is assembled, the lid 54 is coupled with circuit board 12 and chassis body 14 to form the subcircuit cavities. The walls 67 contact board 12 around the defined boundaries of the subcircuits. The walls 67 generally coincide with the ground isolation paths 28. Each of the subcircuits is 5 thereby separately contained and isolated on a single board and using a single lid and chassis body.

A number of clearance areas may also be provided on the bottom side 62 of the chassis lid 54. For example, a low power gain component clearance area 70 may be provided to allow clearance for tall components in the low power gain 10 subcircuit 26. Further, high power gain component clearance areas 72 may be provided to allow clearance for tall components in the high power gain subcircuit cavity 24', monitor and control component clearance areas 74 may be provided to allow clearance for tall components in the monitor and control subcircuit cavity 22', and power supply component clearance areas 76 may be provided to allow for 15 clearance of tall components in the power supply subcircuit cavity 18'. In one embodiment, the clearance areas are set off by approximately 30 mils from a main surface 78 of the bottom side 62 of the inner lid 54, though greater or lesser clearances may be utilized depending on the size of the component to be cleared.

In one embodiment, to further enhance the electrical isolation of the cavities, 20 a soft, electrically-conductive, form-in-place gasket is dispensed on internal chassis lid walls 64 to take up gaps between the internal chassis lid walls 64 and the printed circuit board 12 and paths 28 where the walls and board contact. These gaps may not be tightly controlled, resulting from manufacturing mechanical tolerances in the printed circuit board 12, the chassis lid 54, and the chassis body 14. In one 25 embodiment, the chassis lid walls 64 are shortened by about 35 mils to allow room

for a 65 mil gasket, which is then compressed to ensure a tight EMI shield from one subcircuit cavity to the next when the lid and chassis body are coupled together.

Turning now to Figure 4A, a cutaway side view of a printed circuit board 12 of one embodiment of the present invention is shown. The cutaway side view in Figure 5 4A is taken along the cutaway lines "B" in Figure 1A. Likewise, the cutaway side view in Figure 4B is taken along the cutaway lines "C" in Figure 1B. In one embodiment of the present invention, a four-layer design is utilized to produce a multilayer, printed circuit board. In other embodiments, circuit boards having more or fewer layers than four are possible. For example, two-layer circuit boards or six-10 layer circuit boards may be used in some embodiments.

In the embodiments shown in Figures 4A and 4B, the first conductive layer 16 and a second conductive layer 80 of the printed circuit board 12 are on opposite sides of a first dielectric layer 86. In one embodiment, the second conductive layer 80 and a third conductive layer 82 are separated by a second dielectric layer 88. 15 Similarly, the third conductive layer 82 and a fourth conductive layer 84 are on opposite sides of a third dielectric layer 90.

In one embodiment, the first dielectric layer 86 is a controlled-dielectric manufactured by the Rogers Corporation, model number RO-4350, the third dielectric layer 90 is a dielectric manufactured by the Isola Corporation such as 20 model numbers FR406 or FR408 and the conductive layers are copper. The second dielectric layer 88 is FR-4 prepreg, used for laminating other materials together. Copper clad FR406 or FR408 has a coefficient of thermal expansion that is compatible with RO-4350 dielectric, and the two materials are designed to have similar environmental and processing characteristics so that they may be laminated 25 together for long-term application as a multilayer board. It is also contemplated that

the third dielectric layer 90 be manufactured of Rogers Corporation RO-4350 material.

The cutaway of Figure 4A shows a section of the printed circuit board 12 where the first conductive layer 16 is provided with a break for forming multiple conductive leads. The cutaway of Figure 4B shows a section of the printed circuit board 12 where the first conductive layer 16 is continuous for forming a single lead. Also shown in Figure 4A and 4B are edge plated areas 92, where conductive material is plated through the entirety of the vertical dimension of the cutaway area to provide sufficient grounding interconnect between the layers from the chassis.

The first conductive layer 16 is populated with surface mount components and contains controlled impedance circuits necessary for proper operation of certain components. The layer 16 includes various RF signal paths for the components. The first conductive layer 16 also contains ground planes which act as shields, in conjunction with the internal chassis lid walls 64 for minimization of interference between the various subcircuits on the first conductive layer 16. The second conductive layer 80 is a ground plane, which completes the controlled impedance circuits. The ground plane also acts as a shield, for minimization of interference between the first layer 16 signals and third layer 82 signals. The third conductive layer 82 distributes various signals, such as conditioned voltage and monitoring signals through the printed circuit board. Finally, the fourth conductive layer 84 is nearly completely metallized, and also acts as a ground plane to transfer ground from the chassis body 14 to the printed circuit board 12.

A plurality of plated vias 30, as shown in Figure 1A, are liberally dispersed around the four conductive layers 16, 80, 82, and 84. The plated vias 30 are copper-plated therearound and used to carry signals or ground from layer to layer. When

used for grounding, the vias 30, help ensure a proper ground at all layers by providing connections between all of the ground planes of the first, second, third, and fourth conductive layers 16, 80, 82, and 84. Further, vias 30 are aligned in various patterns as shown in Figure 1A to form the ground isolation paths 28 which, in 5 conjunction with the internal chassis lid walls 64, serve to electrically isolate the subcircuits on the printed circuit board 12. The plated vias 30 forming isolation paths 28 are positioned such that the internal chassis lid walls 64 make electrical contact with the plated vias for desirable grounding characteristics.

To further ensure a proper ground level at all layers, the printed circuit board 10 12 includes a perimeter 94 that is also plated. The perimeter 94 may be plated with copper or another conductive material. The perimeter 94 thus brings ground up from the main surface 44 of the chassis base 14 to each conductive layer 16, 80, 82, and 84 of the printed circuit board 12. The presence of ground planes throughout the printed circuit board 12, in conjunction with plated vias 30 and the internal chassis lid 15 walls 64, provides the desired shielding of the components which reduces the ability of incompatible signals to corrupt each other.

In one embodiment of the present invention, board layout and circuit design are performed concurrently so that signals may be routed without interruption. Further, all circuits are preferably designed to operate on a common board material, 20 such as board material model number RO-4350 manufactured by the Rogers Corporation. According to one embodiment, ground planes are necessary in certain areas for circuit functionality, particularly in circuits with controlled impedances such as in amplification sections and an output power coupler feeding the output power detector subcircuit 32. Functionality of an amplifier may be improved under the 25 present invention because certain circuits, particularly monitor circuits, may be

spread to functionally different areas of the printed circuit board 12. In one embodiment, interconnection devices, such as connectors, jumpers, and discrete wires, are eliminated from the printed circuit board 12. Under the present invention, it is possible for one printed circuit board 12 to receive external control data, control 5 internal subcircuits, monitor subcircuit health, and transmit health status over an external line without the addition of discrete wires to distribute circuit signals.

While the present invention has been described with reference to one or more particular embodiments, those skilled in the art will recognize that many changes may be made thereto without departing from the spirit and scope of the present 10 invention. Each of these embodiments and obvious variations thereof is contemplated as falling within the spirit and scope of the claimed invention, which is set forth in the following claims.